

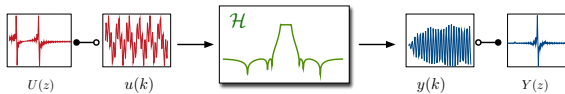
FiXiF toolbox: numerically reliable digital filters

Anastasia Volkova

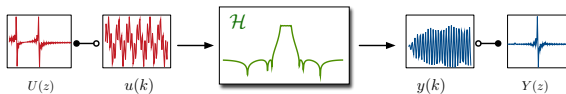
SCAN 2018
September 12, 2018



Linear digital filters



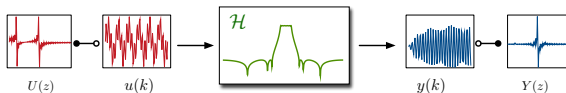
Linear digital filters



Frequency domain

$$H(z) = \frac{0.0495329964 - 0.148598989z^{-2} + \dots}{1 - 2.12060288z^{-1} + 2.7247492z^{-2} + \dots}$$

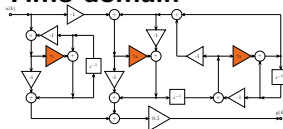
Linear digital filters



Frequency domain

$$H(z) = \frac{0.0495329964 - 0.148598989z^{-2} + \dots}{1 - 2.12060288z^{-1} + 2.7247492z^{-2} + \dots}$$

Time domain

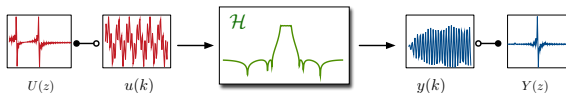


$$\gamma_1 = 0.347586468864209$$

$$\gamma_2 = 0.334748427563068$$

$$\gamma_3 = 0.084938546237853$$

Linear digital filters



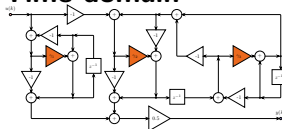
Frequency domain

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coefficient quantization
in Fixed-Point arithmetic

Finite-precision
arithmetic operations

Time domain

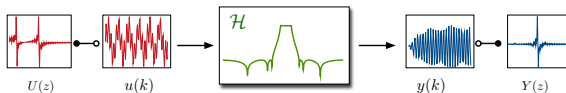


$$\gamma_1 = 0.34765625$$

$$\gamma_2 = 0.3359375$$

$$\gamma_3 = 0.0859375$$

Linear digital filters



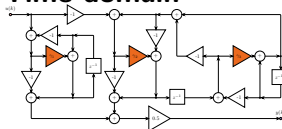
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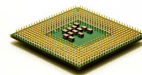
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Circuit / Code

Constraints w.r.t. surface, speed, accuracy ...



Reliable and accurate digital filters

Majority of applications

- do not need numerical guarantee
- well-studied subject
- straightforward implementations
- commercial and free tools



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Safety-critical applications

- require numerical guarantees
- have high(er) cost
- limited literature
- require expert knowledge from engineers
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Reliable and accurate digital filters

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Safety-critical applications

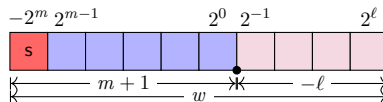
- require numerical guarantees
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Context: safety-critical applications

Goal: generate Fixed-Point codes reliable by construction

Means: interval, floating-point, multiple precision arithmetics

Fixed-Point Arithmetic



w	wordlength
m	Most Significant Bit
ℓ	Least Significant Bit

Problem setting for MSB/LSB choice:

- fix wordlength
- no overflow
- maximize the precision
- bound on the output error

Example: a 15th order lowpass filter

Algorithm	# coefficients
State-Space (canonical)	31
State-Space (balanced)	256
Direct Form II (transposed)	31
Li-Gevers-Sun	102

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Coefficient quantization

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Coefficient quantization

Rounding errors

Existing approaches on FxP implementation

- simulations [Matlab], [D. Báez-López, 2001]
 - non-exhaustive
- noise propagation models [Menard, 2008]
 - does not give *intervals* for outputs
- interval arithmetic [Carreras, 1999], [Vakili, 2013]
 - wrapping effect for recursive systems
- affine arithmetic [Puschel, 2003], [Constantinides, 2006]
 - need to bound ℓ_∞ norm of the filter's output
 - static unroll of the loops



FIXIF

TOOLBOX

"Digital filters reliable by construction"

Available at <https://github.com/fixif>

Joint work with

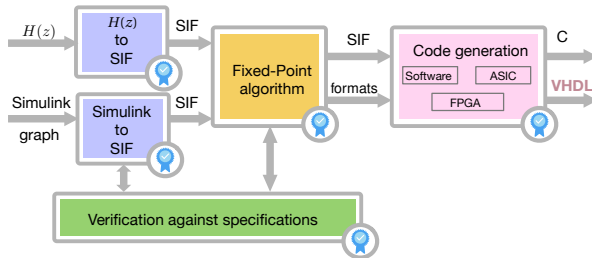


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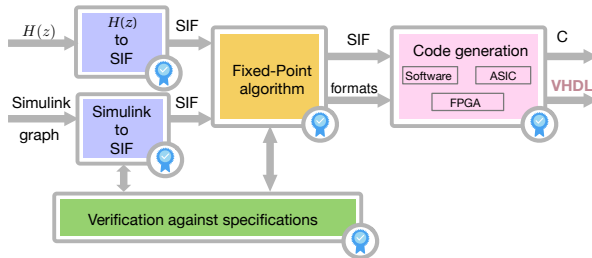


Christoph Lauter
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Compiler overview

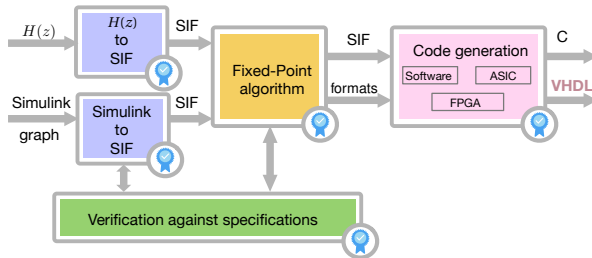


Compiler overview



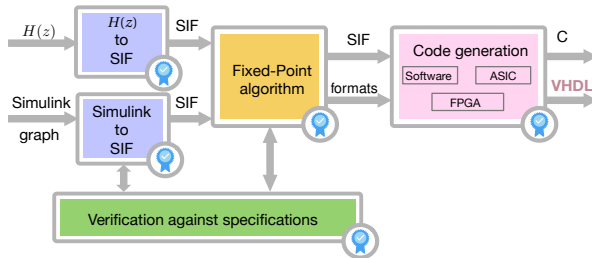
- Internal representation: SIF (matrix-based description)

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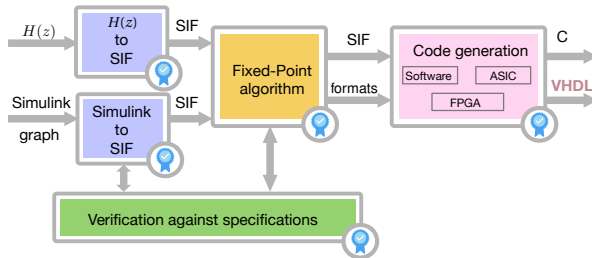
- Internal representation: SIF (matrix-based description)
- Rounding errors: unified, reliable and fast FxP algorithm generation

Compiler overview



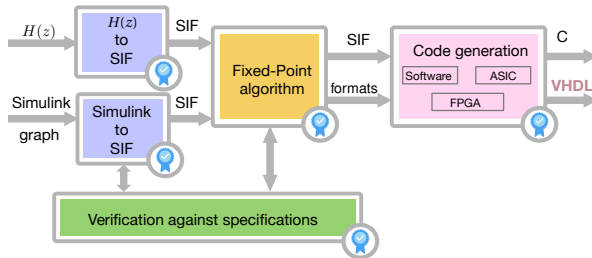
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Compiler overview



- Internal representation: SIF (matrix-based description)
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- Software implementation: floating- and fixed-point C

Compiler overview



- Internal representation: SIF (matrix-based description)
- Rounding errors: unified, reliable and fast FxP algorithm generation
- Quantization errors: rigorous verification against frequency specs
- Software implementation: floating- and fixed-point C
- Hardware implementation: VHDL generation based on FloPoCo^a

^a<http://flopoco.gforge.inria.fr/>

FiXiF tool: under the hood

Front-end

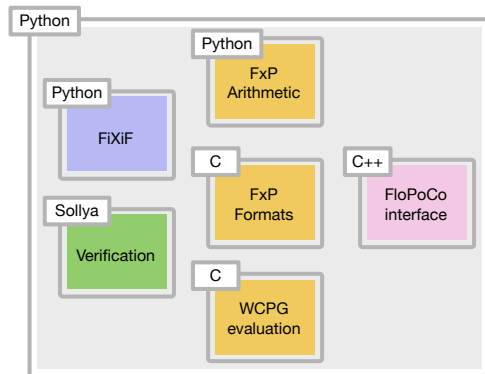
- Python
- Collection of classes

Back-end

- C/C++ libraries
- Sollya procedures
- FloPoCo tool

Requirements:

- MPFR, MPFI, LAPACK
- mpmath



FiXiF – 1

- Filter specifications and transfer function design

```
from fixif import *  
g = Gabarit(48000, [(0, 9600), (12000, None)], [(0.95, 1.05), -20])  
H = g.to_dTF(ftype="butter", method="scipy")  
F = Filter(tf=H)
```

FiXiF – 1

◦ Filter specifications and transfer function design

```
from fixif import *
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```

Type: lowpass (Fs=48000Hz)

Freq. [0Hz,9600Hz]: Passband in [0.95dB, 1.05dB]

Freq. [12000Hz,24000.0Hz]: Stopband at -20dB

$$H(z) = \frac{4.58056194e-05 + 6.87084291e-04 z^{-1} + 4.8095900e-03 z^{-2} + \dots}{1.0 + -1.6206385z^{-1} + 3.20872535z^{-2} + \dots}$$

FiXiF – 1

- Filter specifications and transfer function design

```
from fixif import *  
g = Gabarit(48000, [(0, 9600), (12000, None)], [(0.95, 1.05), -20])  
H = g.to_dTF(fdtype="butter", method="scipy")  
F = Filter(tf=H)
```

- Filter realization (algorithm)

```
R_SS = State_Space(Filter(tf=H))  
R_SS_balanced = State_Space(Filter(tf=H), form="balanced")  
R_DFII = DFII(Filter(tf=H), transposed=True)  
R_LGS = LGS(Filter(tf=H), transposed=True)
```

State-Space and an error model

Take a State-Space algorithm

$$\mathcal{H} \quad \left\{ \begin{array}{lcl} \mathbf{x} \ (k+1) & = & \mathbf{A}\mathbf{x} \ (k) + \mathbf{B}\mathbf{u}(k) \\ \mathbf{y} \ (k) & = & \mathbf{C}\mathbf{x} \ (k) + \mathbf{D}\mathbf{u}(k) \end{array} \right.$$

State-Space and an error model

Take a State-Space algorithm and add rounding

$$\mathcal{H}^\diamond \begin{cases} \mathbf{x}^\diamond(k+1) &= \diamond_{m_x}(\mathbf{A}\mathbf{x}^\diamond(k) + \mathbf{B}\mathbf{u}(k)) \\ \mathbf{y}^\diamond(k) &= \diamond_{m_y}(\mathbf{C}\mathbf{x}^\diamond(k) + \mathbf{D}\mathbf{u}(k)) \end{cases}$$

where $\diamond_m$ is some operator ensuring faithful rounding:

$$|\diamond_m(x) - x| \leq 2^\ell$$

State-Space and an error model

Take a State-Space algorithm and add rounding

$$\mathcal{H}^\diamond \left\{ \begin{array}{lcl} \mathbf{x}^\diamond(k+1) & = & \mathbf{A}\mathbf{x}^\diamond(k) + \mathbf{B}\mathbf{u}(k) + \boldsymbol{\varepsilon}_x(k) \\ \mathbf{y}^\diamond(k) & = & \mathbf{C}\mathbf{x}^\diamond(k) + \mathbf{D}\mathbf{u}(k) + \boldsymbol{\varepsilon}_y(k) \end{array} \right.$$

with

$$|\boldsymbol{\varepsilon}_x(k)| \leq 2^{\ell_x} \quad \text{and} \quad |\boldsymbol{\varepsilon}_y(k)| \leq 2^{\ell_y}$$

State-Space and an error model

Take a State-Space algorithm and add rounding

$$\mathcal{H}^\diamond \begin{cases} \mathbf{x}^\diamond(k+1) &= \mathbf{A}\mathbf{x}^\diamond(k) + \mathbf{B}\mathbf{u}(k) + \boldsymbol{\varepsilon}_x(k) \\ \mathbf{y}^\diamond(k) &= \mathbf{C}\mathbf{x}^\diamond(k) + \mathbf{D}\mathbf{u}(k) + \boldsymbol{\varepsilon}_y(k) \end{cases}$$

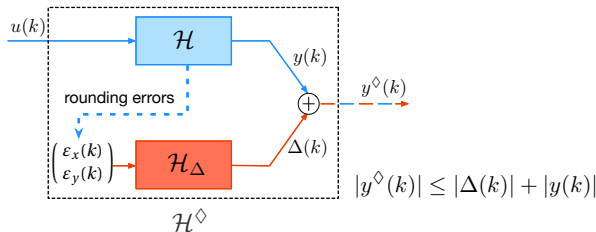
with

$$|\boldsymbol{\varepsilon}_x(k)| \leq 2^{\ell_x}$$

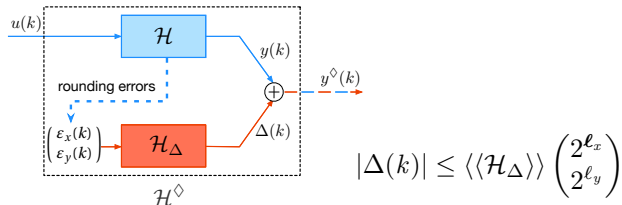
and

$$|\boldsymbol{\varepsilon}_y(k)| \leq 2^{\ell_y}$$

We express $\mathbf{x}(k) - \mathbf{x}^\diamond$ and $\mathbf{y}(k) - \mathbf{y}^\diamond(k)$... using a new filter:



Reliable implementation



Basic bricks:

- Reliable bound on a linear filter's output
 - Worst-Case Peak Gain measure [V-Lauter-Hilaire'15]
- Take into account error propagation
 - Iterative refinement of MSB [V-Lauter-Hilaire'16]
 - Never overestimate MSB by more than one

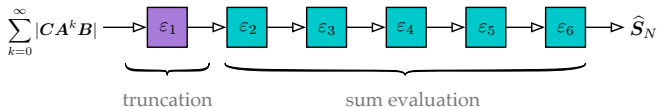
Worst-Case Peak Gain

- Matrix $\langle\langle S \rangle\rangle = \sum_{k=0}^{\infty} |CA^k B|$ [Balakrishnyan-Boyd'91]

Problem: approximation and evaluation of the WCPG with an *a priori* absolute error bound

$$\left| \sum_{k=0}^{\infty} |CA^k B| - \hat{S}_N \right| < \varepsilon$$

Solution [V-Hilaire-Lauter'15]:



direct formula

adaptation of internal precision
s.t. error bound ε_i is satisfied a priori

Techniques: a priori floating-point error analysis, verified inclusions for eigendecomposition, Gershgorin circles computation

FiXiF – 2

Determine the Fixed-Point Formats

```
w = 16
while True:
    w = w - 1
    msb, lsb, error, additionalSteps = FXPF_ABCD(R_SS.A, R_SS.B,
    R_SS.C, R_SS.D, 1.0, w)
```

FiXiF – 2

Determine the Fixed-Point Formats

```
w = 16
while True:
    w = w - 1
    msb, lsb, error, additionalSteps = FXPF_ABCD(R_SS.A, R_SS.B,
    R_SS.C, R_SS.D, 1.0, w)
```

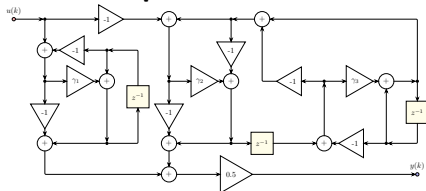
```
Wordlength=[16 16 16 16 16 16 16 16 16 16 16 16 16 16 16 16]
Additional steps: 1
LSB: [-13 -13 -14 -14 -15 -15 -15 -15 -15 -15 -15 -15 -15 -15 -15 -15]
Errors: [[ 0.00086691 0.00121799 0.00152309 0.00172836 0.00158205
0.00121137 0.00079471 0.00048683 0.00029988 0.0001935 0.00013235
0.00009573 0.00007329 0.00005811 0.00004509 0.00152433]]
```

```
:
```

```
Wordlength=[5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5]
Error: LSB reached point when it is larger than initial MSB estimation
Error determining Fixed-Point Formats.
```

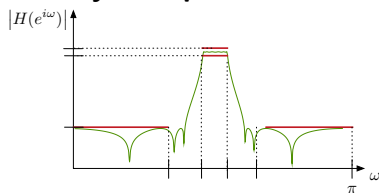
Verification of specifications

Implementation :



$$\gamma_1 = 89 \cdot 2^{-8}, \gamma_2 = 43 \cdot 2^{-7}, \gamma_3 = 11 \cdot 2^{-7}$$

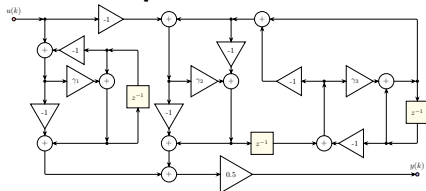
Verify the specifications:



no false positives

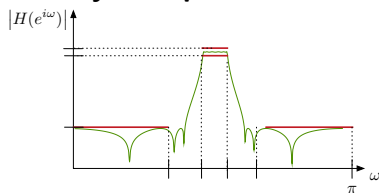
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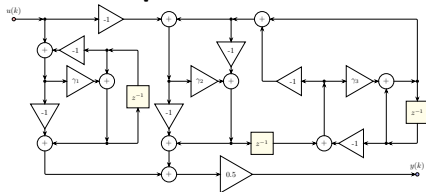
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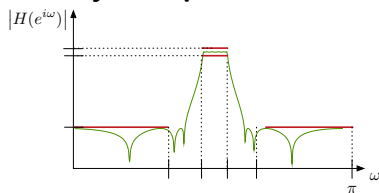
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Solution [V-Lauter-Hilaire'17]

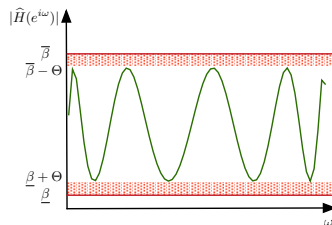
- MP approximation of $\widehat{H}(z)$
 → error bound [Balakrishnyan'92]:

$$\left| (H - \widehat{H})(e^{i\omega}) \right| \leq \Theta$$
- Verification of $\widehat{H}(z)$
 → proof of non-negativity of a polynomial

Verify the specifications:



no false positives



FiXiF – 3

Example of the verification

```
#g = Gabarit(48000, [(0, 9600), (12000, None)], [(0.95, 1.05), -20])  
CheckIfRealizationInGabarit(g, R_SS.quantize(16))
```

FiXiF – 3

Example of the verification

```
#g = Gabarit(48000, [(0, 9600), (12000, None)], [(0.95, 1.05), -20])
CheckIfRealizationInGabarit(g, R_SS.quantize(16))
```

Overall check okay: false

Computing this result took 2258ms

The following issues have been found:

Issue in subdomain $\Omega = \pi * [0; 0.4]$ at $\omega = \pi * [0; 5.3172964416593691658936355013934529087741417880979e-53]$:
 $|H(\exp(j*\omega))|$ should be between $10^{4.75e-2}$ and $10^{(1.05 / 20)}$ but
 evaluates to $[1.0000497795523700519233990284759435759640164376565;$
 $1.0000497795523700519233990284759435759640164376566] =$
 $10^{([4.323689366417695635232485092238116632596970479257e-4;$
 $4.3236893664176956352324850922381166325969704866858e-4] / 20)}$

⋮

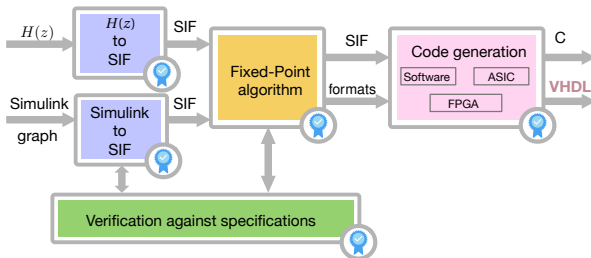
Behind the scenes

During these 2.26 seconds...

- Rational arithmetic
 - proof of non-negativity of a real polynomial
- Interval arithmetic
 - Verified Inclusions for the eigenvalues
- Dynamic Multiple-Precision arithmetic
 - Computation of Gershgorin circles
 - Evaluation of WCPG
- Floating-Point arithmetic
 - Computation of eigenvalues with LAPACK
 - Matrix arithmetic with *a priori* error bounds

...to verify an implementation in Fixed-Point arithmetic!

Back to the generator



- Rigorous use of a combination of different arithmetics
- Numerical guarantees in time and frequency domains
- Unifying internal representation
- Easily extendable modular implementation

Beyond FiXiF

Verification of frequency specifications:

- software-hardware co-design of FIR filters¹

Worst-Case Peak Gain evaluation:

- digital filters computed just right on FPGAs²

Internal representation SIF

- error analysis for multiplierless FFT³

Formal proofs of the code generator

- Coq proofs of the WCPG algorithm⁴

¹with M. Kumm (Kassel University) and S. Filip (Inria Rennes)

²with F. de Dinechin (INSA Lyon) and M. Istoan (Imperial College)

³with F. Qureshi and J. Takala (Tampere University)

⁴with S. Boldo and D. Gallois (Inria Paris)

FIXIF

TOOLBOX

"Digital filters reliable by construction"

<https://github.com/fixif>

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Verification of a transfer function

We need to show that $\forall z = e^{i\omega}, \omega \in \Omega \subseteq [0, \pi]$

$$\underline{\beta} \leq |H(z)| \leq \overline{\beta}$$

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$$\underline{\beta} \leq |H(z)| \leq \overline{\beta}$$

...which boils down to showing that $\forall \xi \in \Xi \subseteq [0, 1]$

$$f(\xi) \geq 0$$

with $f \in \mathbb{R}[\xi]$.

Verification that f is positive or zero

Our algorithm is based on:

- the Sturm's technique that gives the number of real roots of a polynomial
- a subdivision on sub-intervals
- evaluations in Multiple-Precision Interval Arithmetic